

Uncertainty Analysis for Four Terminal-Pair Capacitance and Dissipation Factor Characterization at 1 and 10 MHz

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Abstract—The Electricity Division at the National Institute of Standards and Technology (NIST, formerly NBS) has developed the capability to characterize capacitance and dissipation factor for four terminal-pair (4TP) air dielectric capacitors at frequencies from 1 kHz to 10 MHz. The method, based on work by Cutkosky and Jones of NBS and recent developments by Hewlett-Packard Japan, involves single-port network analyzer impedance measurements at frequencies from 40 MHz to 200 MHz and capacitance measurements using a precision 1 kHz capacitance meter. A mathematical regression algorithm is used to extrapolate inductance and resistance measurements from the network analyzer down to 1 kHz in order to predict capacitance and dissipation factor from 1 kHz to 10 MHz. A comprehensive uncertainty analysis for the procedure is presented.

Index Terms—Capacitance, dissipation factor, four terminal pair, impedance, precision measurements, uncertainty analysis.

I. INTRODUCTION

THE Electricity Division at the National Institute of Standards and Technology (NIST, formerly NBS) has implemented a system to characterize capacitance and dissipation factor for four terminal-pair (4TP) air dielectric capacitors at frequencies from 1 kHz to 10 MHz [1]. The method is based on work by Cutkosky [2], [3] and Jones [4], [5] of NBS and recent developments by Yokoi *et al.* [6], [7] as well as Yonekura and Wakasugi [8] of Hewlett-Packard Japan. This paper describes an extensive uncertainty analysis of the measurement system. The analysis has been divided into three areas: 1-kHz capacitance measurements; network analyzer impedance measurements (covering frequencies from 40 to 200 MHz); and a mathematical extrapolation algorithm that regresses the high-frequency characterization down to frequencies of 10 MHz and below [6], [7]. This algorithm is referred to as the capacitor frequency characteristic prediction (CFCP) method. The capacitance and dissipation factor characteristics at 1 and 10 MHz are produced by applying the CFCP algorithm to the 1-kHz capacitance as well as the high-frequency (40–200 MHz) impedance measurements.

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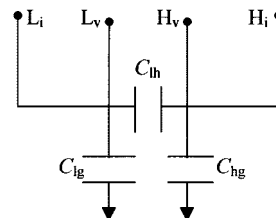


Fig. 1. Four terminal-pair capacitor: simple model.

TABLE I
TYPE A STANDARD UNCERTAINTIES FOR
1-KHz MEASUREMENTS

| Nominal Capacitance (pF) | Measured Parameter | Measured Capacitance (pF) | Type A Uncertainty (ppm) |
|--------------------------|--------------------|---------------------------|--------------------------|
| 1 | C _{lh} | 1.0000142 | 5 |
| | C _{lg} | 48.82658 | 37 |
| | C _{hg} | 17.67269 | 86 |
| 10 | C _{lh} | 10.000745 | 4 |
| | C _{lg} | 29.16876 | 14 |
| | C _{hg} | 25.02418 | 30 |
| 100 | C _{lh} | 99.99892 | 7 |
| | C _{lg} | 32.27685 | 30 |
| | C _{hg} | 31.03866 | 37 |
| 1000 | C _{lh} | 1000.0739 | 5 |
| | C _{lg} | 34.98904 | 164 |
| | C _{hg} | 33.36330 | 263 |

Capacitors characterized using this technique will be used as impedance reference standards for a general-purpose digital impedance bridge recently developed at NIST to calibrate inductors and ac resistors [9]. The technique is also to be employed in a future NIST special test for 4TP capacitance and dissipation factor.

II. 1-KHz CAPACITANCE MEASUREMENT UNCERTAINTY

Fig. 1 shows the simple 4TP capacitor circuit model, where C_{lh} is the low-to-high capacitance and C_{lg} and C_{hg} are low-to-ground and high-to-ground leakage capacitances, respectively. These capacitance components are repeatedly measured over time to establish repeatability using a 1-kHz capacitance meter. Table I presents Type A relative standard uncertainties for 1-kHz measurements of the 1-, 10-, 100-, and 1000-pF standard capacitors. Uncertainties are given as parts in 10^6 and labeled ppm. The Type B relative standard uncertainty for the 1-kHz capacitance meter is about 10 ppm [10].

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| 14. ABSTRACT The Electricity Division at the National Institute of Standards and Technology (NIST, formerly NBS) has developed the capability to characterize capacitance and dissipation factor for four terminal-pair (4TP) air dielectric capacitors at frequencies from 1 kHz to 10 MHz. The method, based on work by Cutkosy and Jones of NBS and recent developments by Hewlett-Packard Japan involves single-port network analyzer impedance measurements at frequencies from 40 MHz to 200 MHz and capacitance measurements using a precision 1 kHz capacitance meter. A mathematical regression algorithm is used to extrapolate inductance and resistance measurements from the network analyzer down to 1 kHz in order to predict capacitance and dissipation factor from 1 kHz to 10 MHz. A comprehensive uncertainty analysis for the procedure is presented. | | | | | |
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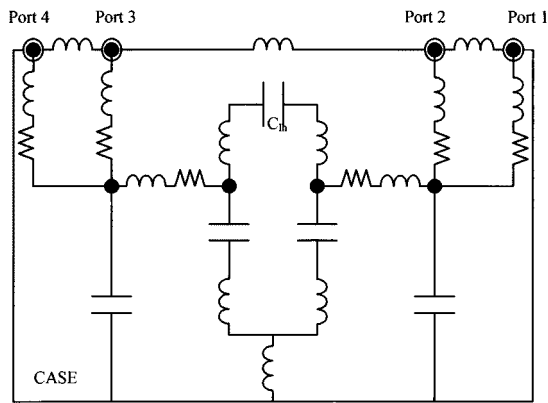


Fig. 2. Yonekura component model.

Since the results of the 1-kHz capacitance measurements are used in the CFCP method, there will be an uncertainty contribution from the 1-kHz measurements to the final capacitance and dissipation factor results. Software simulations were performed to determine the uncertainty components of the 4TP capacitance and dissipation factor due to the 1-kHz capacitance measurements.

In 1990, Yonekura and Wakasugi described a circuit model of 4TP capacitors (of the type described in this paper) [8]. They disassembled several of these capacitors and measured and published the values of typical model components, e.g., connector and lead inductances. The following simulations are based on this model and component values (see Fig. 2).

Equations were developed to describe the Yonekura model in terms of the model components. Using published values from [8], we computed the 4TP capacitance and dissipation, as well as the single-port data needed for the CFCP algorithm for this “reference” capacitor.

The sensitivity of the CFCP method was tested by applying a fixed set of single-port (simulated network analyzer) data, while randomly varying the 1-kHz capacitance data based on the values given in Table I. Predictions of the 4TP capacitance and dissipation factor from the CFCP method were obtained for the multiple simulations at 1 and 10 MHz. The predictions were compared with the exact solution of the reference capacitor to estimate the Type A uncertainty contribution in the final results due to the 1-kHz measurements. The Type B uncertainty simulations were run just as the Type A simulations except that the 1-kHz capacitance values were offset by 10 ppm instead of randomly scattered prior to running the CFCP algorithm. The uncertainties produced from this simulation are reported in Section V as 1-kHz Type A and B standard uncertainties.

Note that the Type B uncertainty of the capacitance meter has been approximated conservatively at 10 ppm.

III. NETWORK ANALYZER MEASUREMENT UNCERTAINTY

The high-frequency single-port measurements used by the CFCP method are provided by a precision network analyzer. Measurements are made from 40 to 200 MHz, depending on the capacitor. The actual measured quantity is the scattering parameter S_{11} , which is converted into impedance. The network analyzer contributes both Type A and B uncertainties. The Type

A component of uncertainty is the standard deviation of capacitance and dissipation factors based on repeated network analyzer measurements applied to the CFCP algorithm. Note that this value also includes random variations of the capacitor.

The Type B component of uncertainty due to the network analyzer was estimated using software simulations. Again, the solution of the Yonekura circuit model [8] was used to compute the true 4TP capacitance and dissipation factor of the reference capacitor at frequencies of 1 and 10 MHz. Ideal S_{11} parameters (that would be generated by an errorless network analyzer) were also computed from the circuit equations and applied to the CFCP algorithm to predict 1- and 10-MHz capacitance and dissipation factor behavior. The calculated S_{11} values were then modified to simulate a network analyzer with offset, gain, and frequency response errors within the manufacturer’s specifications [11]. S_{11} measurements of a NIST-calibrated precision 20-cm air line indicated that the network analyzer was within these specifications.

The Type A and B standard uncertainties in the 4TP characterization due to the network analyzer are reported in Section V.

IV. REGRESSION ALGORITHM UNCERTAINTY DUE TO VARIATIONS IN CAPACITATOR MANUFACTURING

Still more simulations were performed to determine the uncertainty components of the 4TP capacitance and dissipation factor introduced by the CFCP method. The regression algorithm extrapolates the network analyzer impedance measurements (made over a range of frequencies chosen somewhere between 40–200 MHz) down to 10 MHz and below using the 1-kHz capacitance measurement values, described briefly above, as references. The regression parameters were selected to optimally predict the capacitance frequency characteristic for nominal capacitor values. This test was set up to determine the sensitivity of the method to variability in manufacturing of the capacitor standards.

The circuit solution of the Yonekura model provides reference values of capacitance and dissipation factor at the frequencies of interest. It is also used to obtain high-frequency single-port values. Simulated network analyzer measurement data were used to iteratively extrapolate to the frequencies of 1 and 10 MHz with normally distributed random errors injected into the Yonekura model components according to each component’s uncertainty [8]. For each reference capacitor, the exact solution and the value predicted by the CFCP method was compared. The Type B standard uncertainties attributed to the method are reported in Section V.

V. UNCERTAINTY RESULTS

Table II labels the standard uncertainty components, and Table III shows the values of the components as well as the expanded standard uncertainties for capacitance and dissipation factor characterization of the standard 4TP capacitors (1, 10, 100, and 1000 pF) at frequencies of 1 and 10 MHz. The uncertainty components are root-sum-squared and then multiplied by two to produce the expanded standard uncertainty values. All capacitance uncertainty components are given in parts in 10^6 , labeled as ppm, and all dissipation factor uncertainty

TABLE II
UNCERTAINTY COMPONENT DESCRIPTIONS

| | |
|----|---|
| a1 | Type A 1 kHz capacitance standard uncertainty |
| a2 | Type A network analyzer standard uncertainty |
| b1 | Type B 1 kHz capacitance standard uncertainty |
| b2 | Type B network analyzer standard uncertainty |
| b3 | Type B CFCP standard uncertainty |

TABLE III
RELATIVE UNCERTAINTY COMPONENTS AND EXPANDED UNCERTAINTIES
($k = 2$)

| | a1 | a2 | b1 | b2 | b3 | U ($k=2$) |
|---|----|-----|-----|------|------|-------------|
| 1 pF Capacitance (ppm) | | | | | | |
| 1 MHz | 5 | 0 | 10 | 250 | 250 | 700 |
| 10 MHz | 5 | 8 | 10 | 5000 | 5000 | 14000 |
| 1 pF Dissipations Factor (μ rad) | | | | | | |
| 1 MHz | 5 | 5 | 10 | 100 | 100 | 300 |
| 10 MHz | 5 | 153 | 10 | 2000 | 2000 | 5700 |
| 10 pF Capacitance (ppm) | | | | | | |
| 1 MHz | 4 | 1 | 10 | 5 | 1 | 24 |
| 10 MHz | 4 | 76 | 31 | 520 | 75 | 1100 |
| 10 pF Dissipation Factor (μ rad) | | | | | | |
| 1 MHz | 4 | 0 | 10 | 9 | 2 | 28 |
| 10 MHz | 4 | 2 | 10 | 295 | 70 | 610 |
| 100 pF Capacitance (ppm) | | | | | | |
| 1 MHz | 7 | 0 | 10 | 5 | 0 | 26 |
| 10 MHz | 7 | 9 | 10 | 545 | 7 | 1100 |
| 100 pF Dissipation Factor (μ rad) | | | | | | |
| 1 MHz | 7 | 0 | 10 | 1 | 0 | 24 |
| 10 MHz | 7 | 0 | 10 | 51 | 2 | 105 |
| 1000 pF Capacitance (ppm) | | | | | | |
| 1 MHz | 5 | 0 | 10 | 5 | 2 | 25 |
| 10 MHz | 5 | 8 | 36 | 474 | 255 | 1100 |
| 1000 pF Dissipation Factor (μ rad) | | | | | | |
| 1 MHz | 5 | 0 | 22 | 6 | 8 | 50 |
| 10 MHz | 5 | 0 | 610 | 203 | 266 | 1400 |

components are given in microradians, labeled as μ rad. These values will be refined and reevaluated as the authors gain experience with the measurement system.

VI. FUTURE WORK

The uncertainty analysis reported in this paper consists of simulations that determine the statistical variation of the components of the 4TP capacitor characterization technique. Some assumptions are made regarding the errors and the circuit model derived from measurements of a set of standard capacitors. A theoretical analysis of the network analyzer should be performed in order to compare with the simulations performed and reported upon here.

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